



High Aspect Ratio Electron Beam, High Efficiency Interaction Structure, and High Power Amplifier Design*

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DARPA/MTO High frequency Integrated Vacuum Electronics (HiFIVE) Program

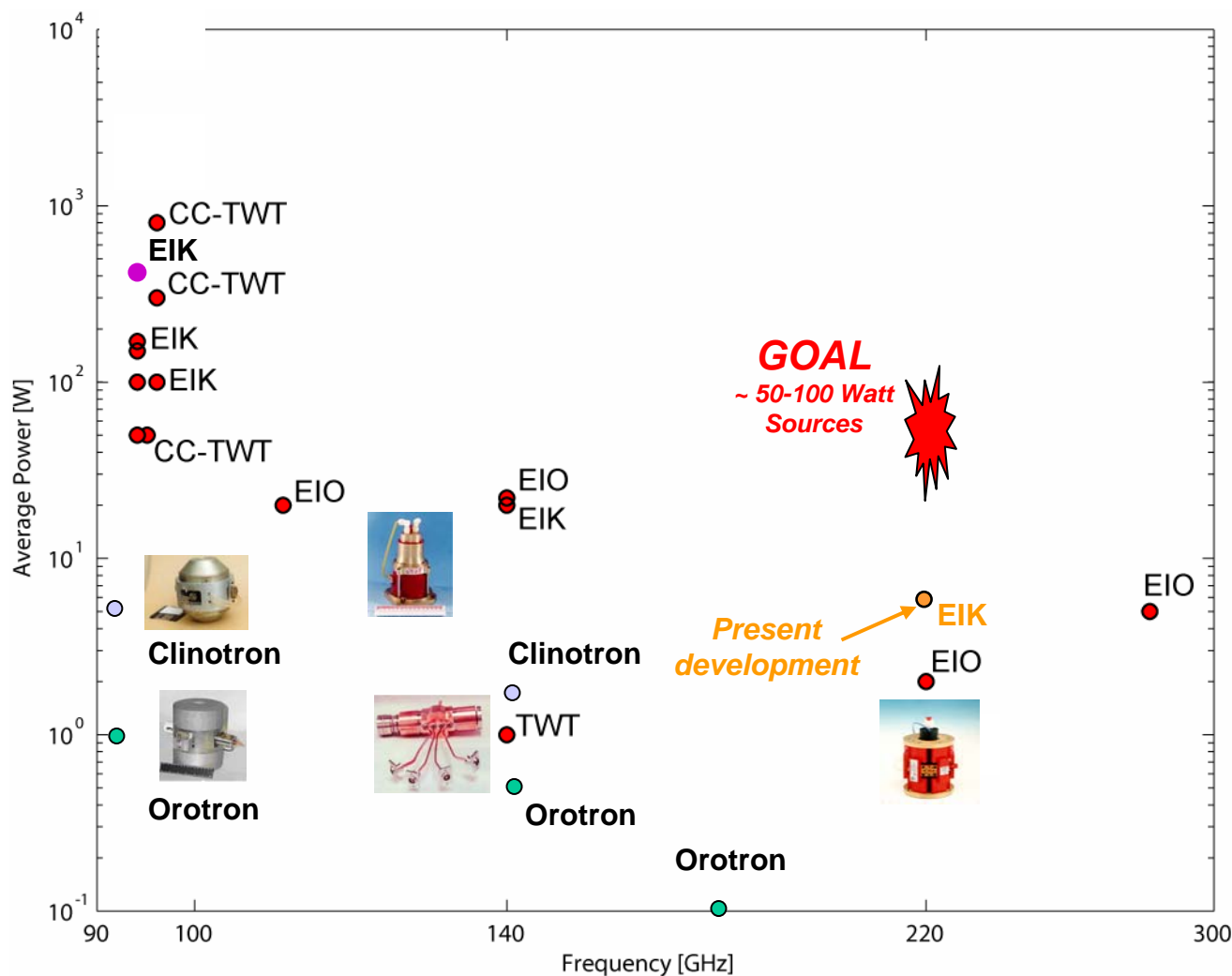
* Work sponsored by DARPA/MTO



High Aspect Ratio Electron Beam and High Efficiency Interaction Structure

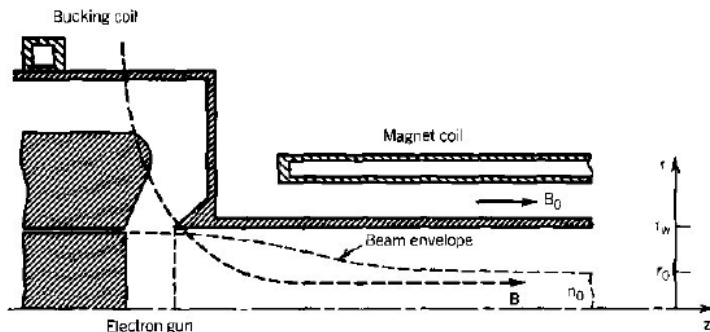


Upper MM-Wave Slow Wave Sources State-of-the-Art





Magnetic Focusing and Transport of Electron Beams – a Key Limiting Factor



From S. Humphries, "Charged Particle Beams"

- **Beam envelope equation:**

$$R'' + k^2 R - \frac{K}{R} - \frac{\varepsilon^2}{R^3} = 0$$

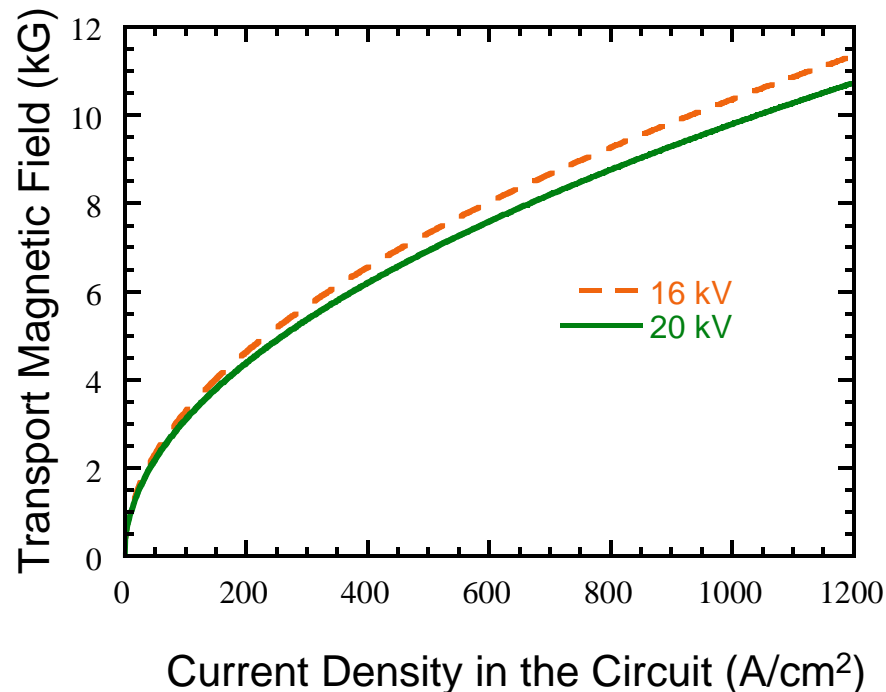
- **Brillouin flow:**

$$k^2 R - \frac{K}{R} = 0 \Rightarrow B_{Brillouin} = 262 \left(\frac{J^{1/2}}{V_b^{1/4}} \right) \text{ Gauss}$$

$$K \equiv \frac{2I}{(\beta\gamma)^3 (17.0 \text{ kA})}$$

- J in A/cm²
- V_b in kV

$$B_{transport} \sim 2.5 \times B_{Brillouin}$$



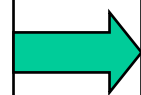


Power vs. Frequency (f) Scaling for Round Beam Devices – Basic Issues



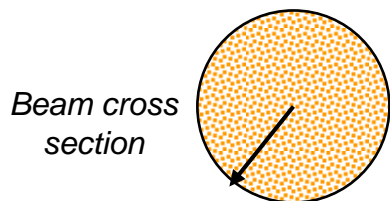
Fixed Upper Limit on Circuit Magnetic Field B_{max}
due to Permanent Magnet Technology (~ 11 kG)

Practical Upper Limit on Beam Voltage Due to
Systems Limitations (~ 20 kV)



Fixed Upper Limit on Current Density
in the Circuit J_{max} (~ 1200 A/cm²)

Low Frequency



High
Frequency



Circuit Radius $\sim 1/f$

Beam Radius $\sim 1/f$

Beam Area $\sim 1/f^2$

Beam Current (I_b) $\sim J_{max}(1/f^2)$

Several Possible Scalings. For Example.....

Circuit not breakdown limited (fixed beam voltage V_b)

Beam Power (P_b) $\sim V_b I_b \sim V_c J_{max} (1/f^2)$

Output Power $\sim \eta P_b \sim (1/f^{0.5}) (1/f^2) \sim (1/f^{5/2})$

Circuit breakdown limited (beam voltage V_b scaling as $1/f$)

Beam Power (P_b) $\sim V_b I_b \sim J_{max} (1/f^3)$

Output Power $\sim \eta P_b \sim (1/f^{0.5}) (1/f^3) \sim (1/f^{7/2})$

Interaction impedance limited

Output Power $\sim \eta R I_b^2 \sim (1/f^{0.5}) (1/f^4) \sim (1/f^{9/2})$



Scaling formulae for traveling wave devices and standing wave devices



Device

Traveling wave

Standing wave

Peak RF Power (W)

$$N \times 24 \left(\frac{1}{f} \right)^{8/3} (V_b)^{13/6} (J)^{4/3}$$

$$N \times 150 \left(\frac{1}{f} \right)^{13/4} (V_b)^{5/2} (J)^{3/2}$$

Peak Elect. Eff. (%)

$$14.4 \left(\frac{1}{f} \right)^{2/3} (V_b)^{1/6} (J)^{1/3}$$

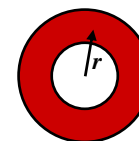
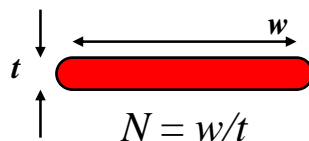
$$48 \left(\frac{1}{f} \right)^{5/4} (V_b)^{1/2} (J)^{1/2}$$

CW RF Power (W)

$$N \times 2.4 \left(\frac{1}{f} \right)^{8/3} (V_b)^{13/6} (J)^{4/3}$$

$$N \times 15 \left(\frac{1}{f} \right)^{13/4} (V_b)^{5/2} (J)^{3/2}$$

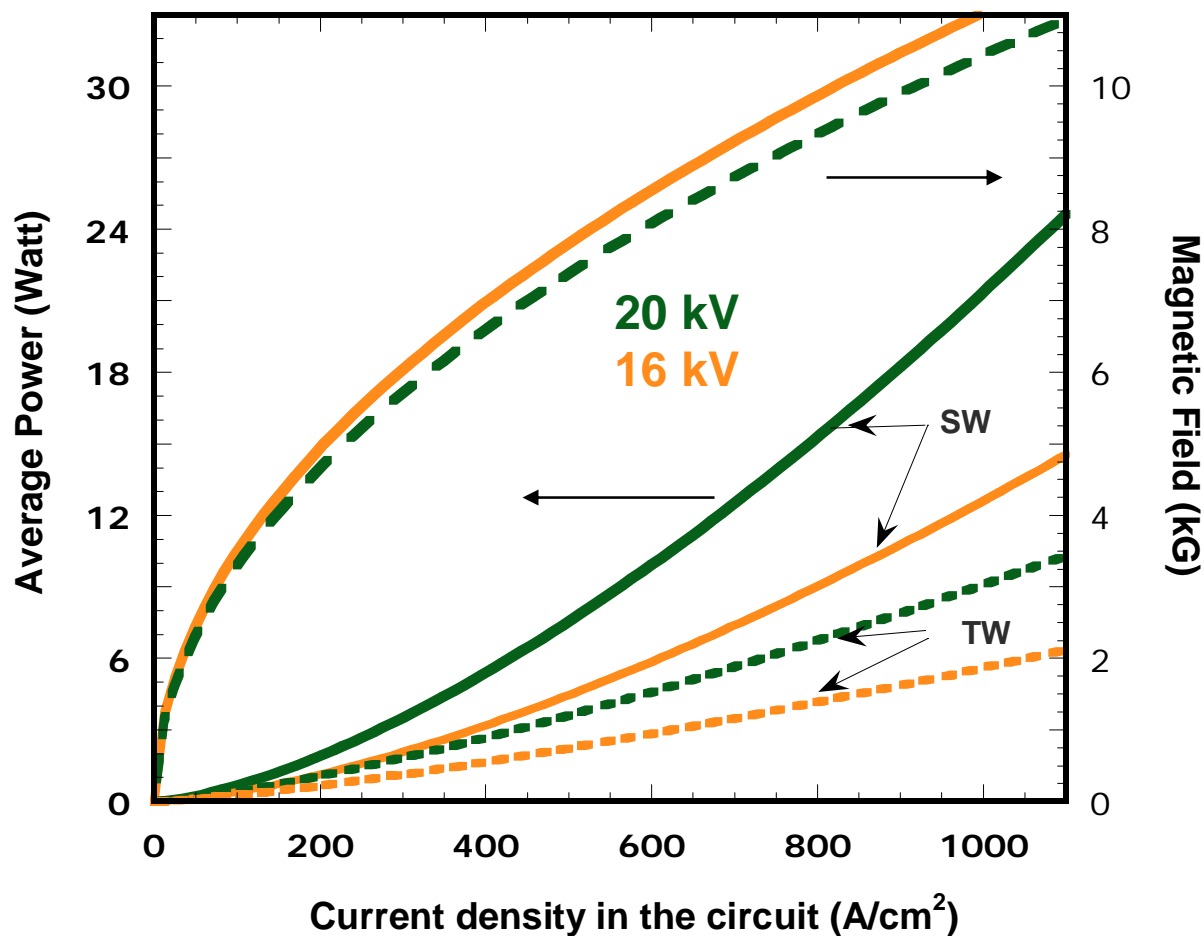
Numeric values from state-of-the art anchor points, f in GHz, V_b in kV, and J in A/cm². N is the beam aspect ratio or number of beams.



$$N = 2\pi r/t$$

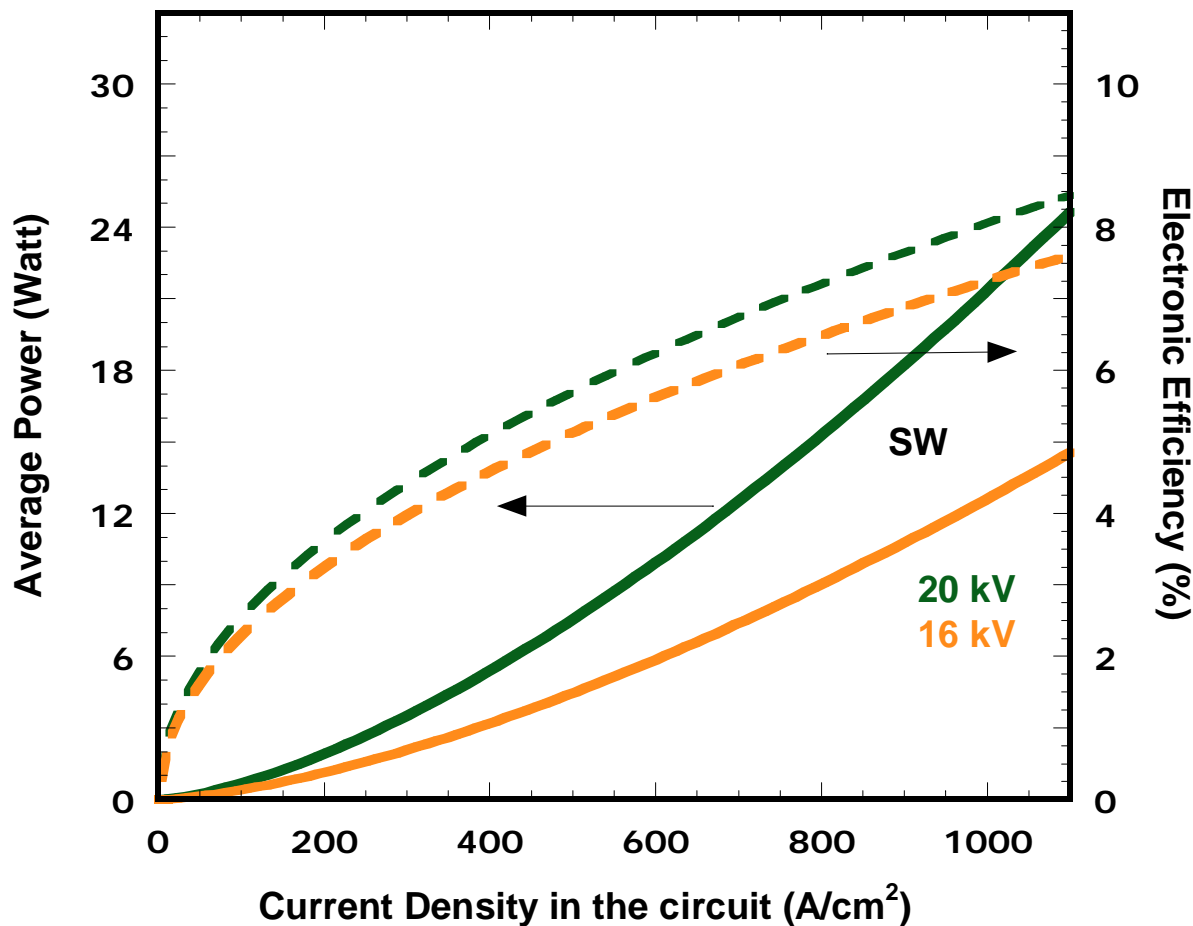


Scaling at 220 GHz





More Scaling at 220 GHz



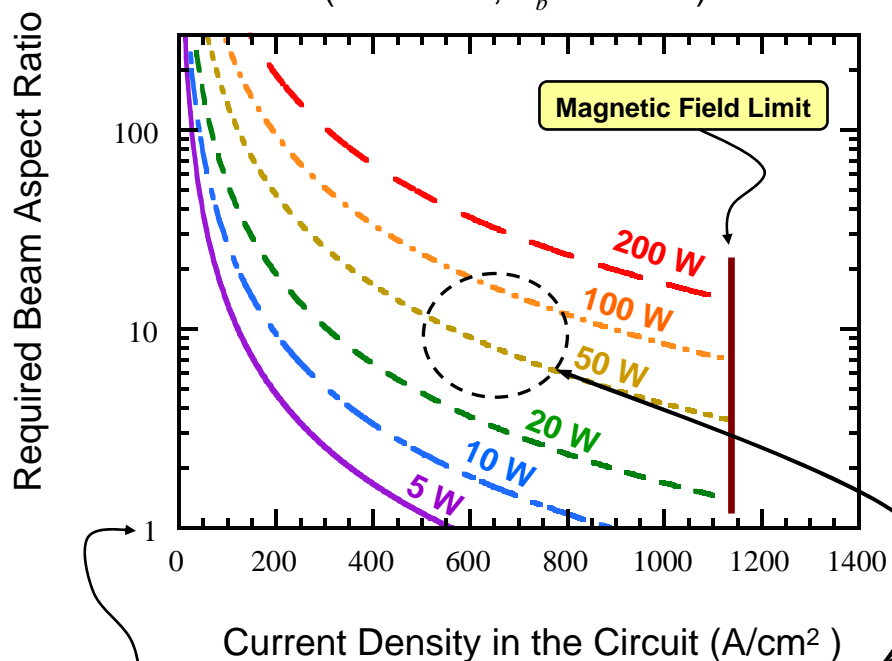


Projected 220 GHz Performance at $V = 16$ kV



SW

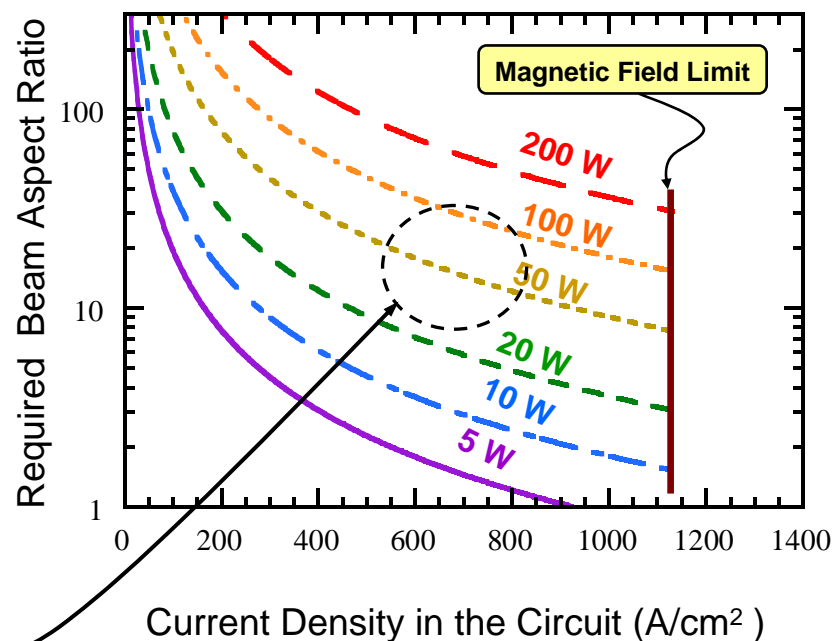
Curves for Various CW Output Powers
(220 GHz, $V_b = 16$ kV)



Round Beam
(Aspect Ratio = 1)

TW

Curves for Various CW Output Powers
(220 GHz, $V_b = 16$ kV)



Current density $\sim 750 A/cm^2$

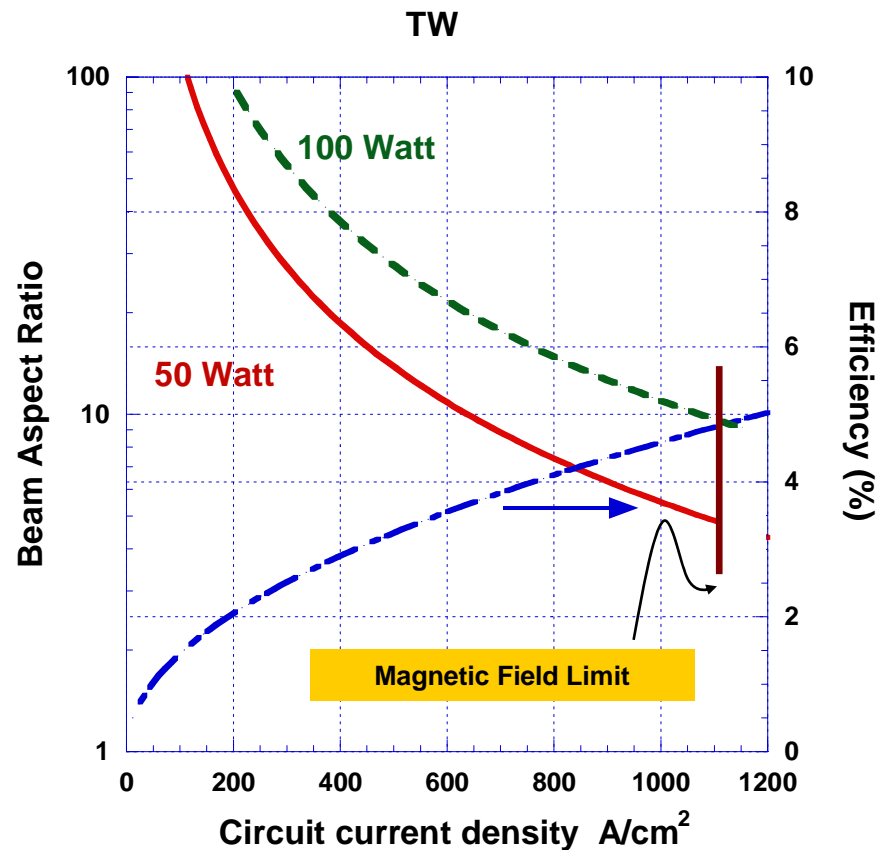
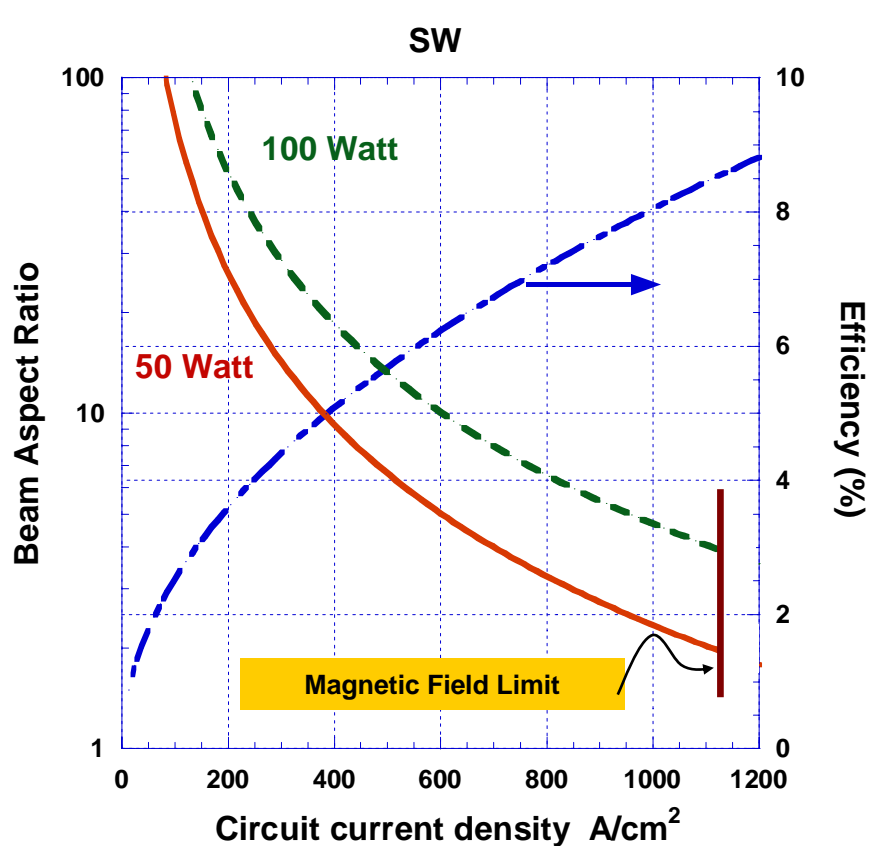
Beam aspect ratio $\sim 10-25$



50 W Output Power at
220 GHz



Projected 220 GHz Performance at $V = 20$ kV





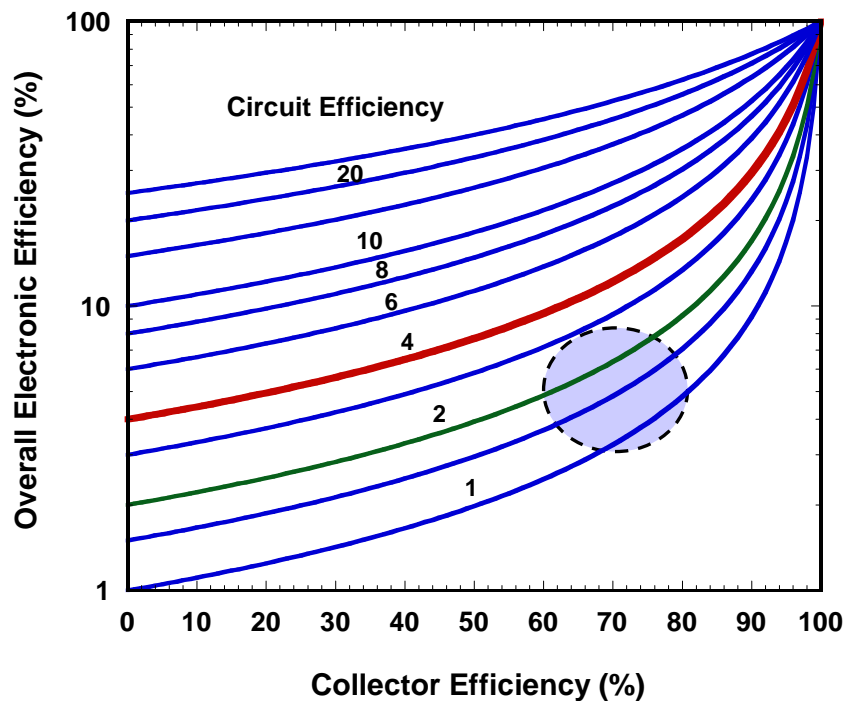
Multi-stage Depressed-Collector Technology for Efficiency Enhancement



Electronic Efficiency

$$\eta_{total} \approx \frac{\eta_{circuit}}{1 - \eta_{collector} (1 - \eta_{circuit})}$$

Multi-stage depressed collector
State-of-the-art efficiency > 80%.





Component Design

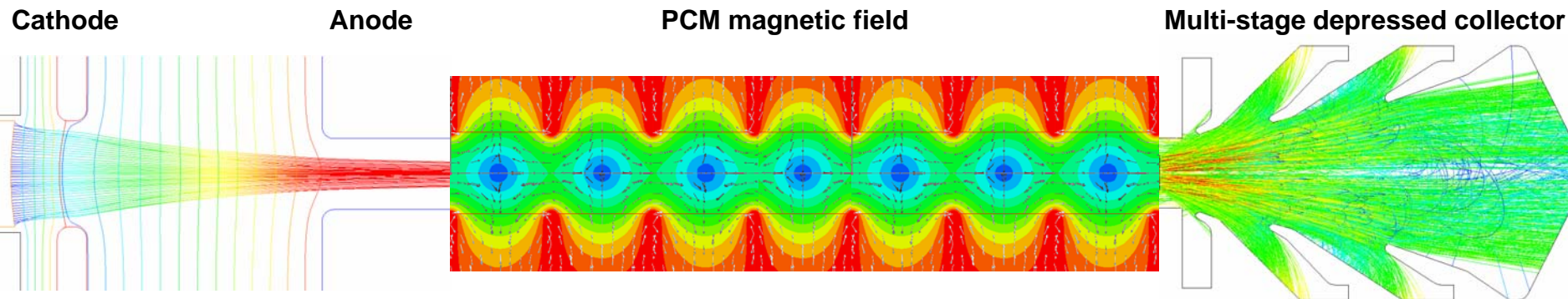


- MICHELLE 3D and Maxwell 3D addressing the beam generation, propagation and collection

Beam formation

Beam transport

Beam energy recovery





Focusing Options with Permanent Magnets



– Periodic Cusp Magnet (PCM) fields

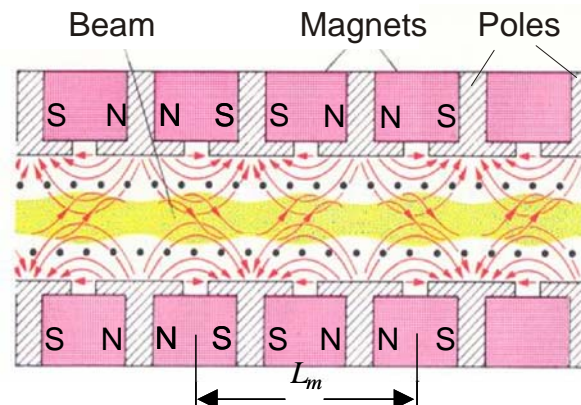
- Very common focusing scheme for conventional tubes

- Challenges

- Obtaining sufficient field
- Transverse plane focusing
- Small parts
- Magnetization homogeneity

- Advantages

- Stable transport for long distances



$$2\omega_p^2 \leq \left(\frac{eB_0}{m} \right)^2$$

$$\ll 16 \frac{e}{m} V \left(\frac{\pi}{L_M} \right)^2$$

– Permanent Magnet (PM) axial (solenoidal) fields

- Used in some compact MMW tubes

- Challenges

- Diocotron instability

- Advantages

- Higher fields obtainable
- Simple magnetic geometry

$$B [kG] > 0.32 \frac{I [A]}{V [kV]} \frac{L [cm]}{A_{beam} [cm^2]}$$

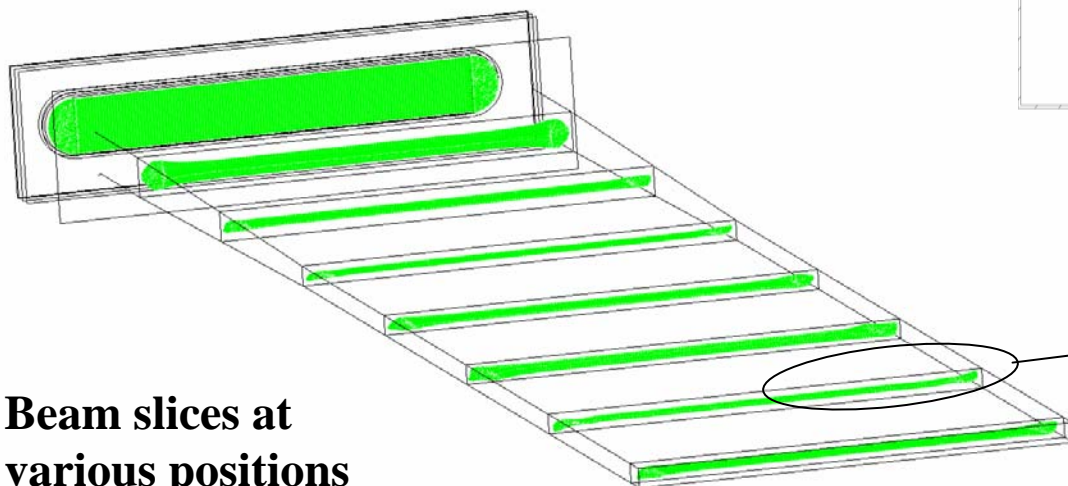
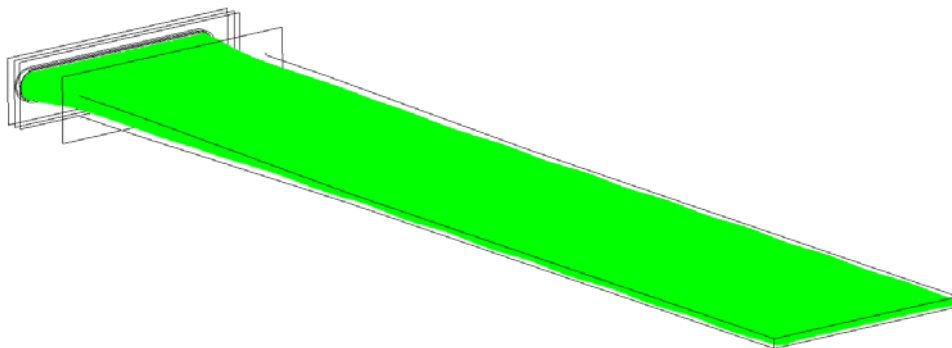
Condition for $L < L_{Diocotron}$



Sheet-Beam Transport in Solenoidal Field

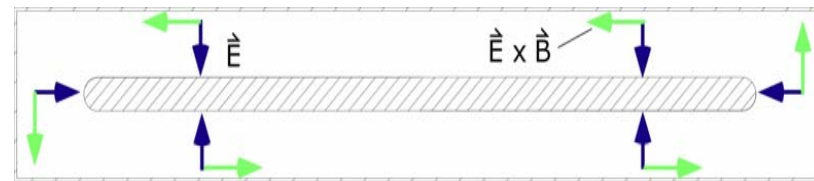


MICHELLE Simulation

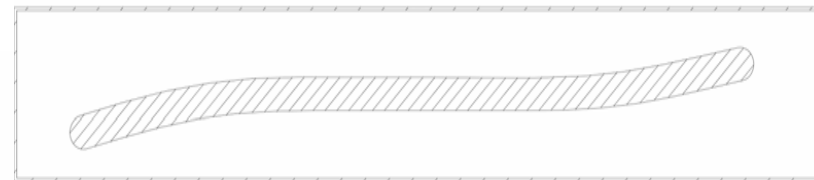


Beam slices at various positions

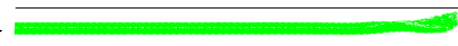
ExB Drift Diocotron Instability



(a)



(b)



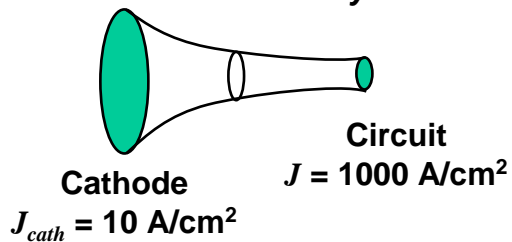


Beam Compression Requirements



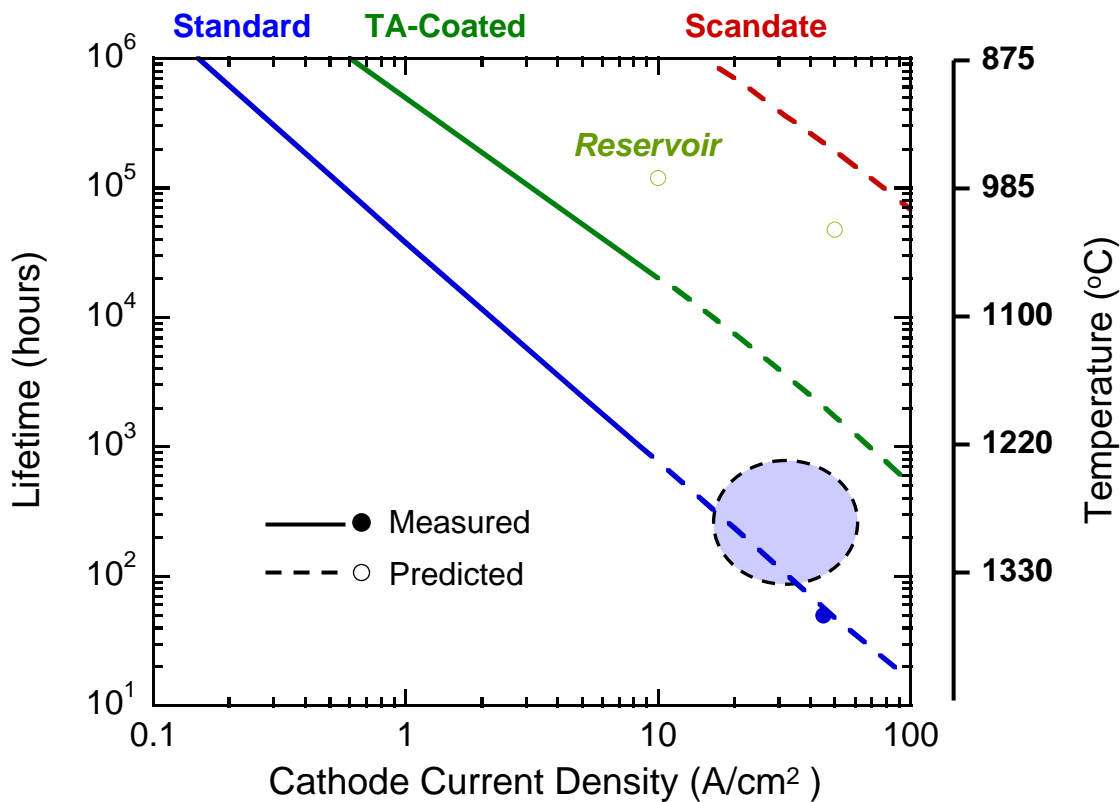
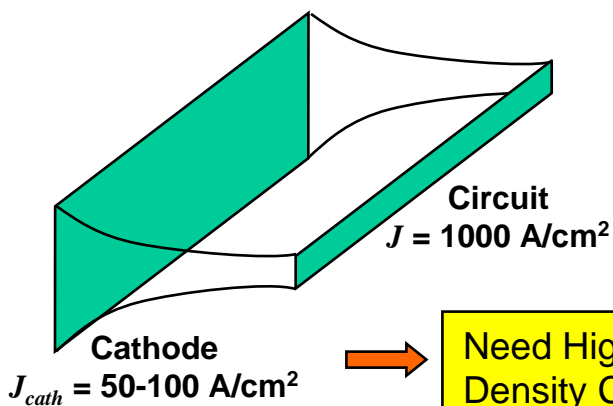
Round Beam – Compression is Two Dimensional

If radius is reduced by a factor of 10



Sheet Beam – Compression is Mainly One Dimensional

If height is reduced by a factor of 10-20





High Power Amplifier Design



High Power Amplifier Design

- Perform theoretical analyses and simulations using physics-based modeling and simulation tools, such as *MICHELLE, CHRISTINE and TESLA (NRL)* and COTS software, such as:

MAXWELL and HFSS (<http://www.ansoft.com/>),

MAGIC (<http://www.magictoolsuite.com/>),

ANALYST (<http://www.staarinc.com/>),

ANSYS (<http://www.ansys.com/>)

- Analyze beam-wave interaction in circuit
 - > Power
 - > Efficiency
 - > Bandwidth
- Determine limitations and devise solutions to key problems
 - > Stability
 - > Breakdown
 - > Thermal
- Design, fabricate, and cold test most promising structures to determine optimum configuration and fabrication techniques



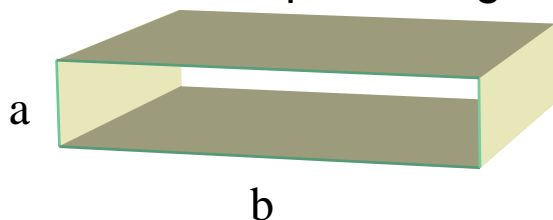
Higher Order Modes in Sheet Beam Devices



- Sheet beam RF structures are inherently overmoded
 - Beam/circuit lateral size becomes large compared to RF wavelength
 - Many competing modes appear with similar resonant frequencies
- Need an understanding of mode competition

**Mode competition is a universal problem
for sheet beam RF amplifiers**

Consider a simple waveguide:



TM mode cutoff frequencies are:

$$\frac{\omega_{m,n}^2}{c^2} = \left(\frac{m\pi}{a} \right)^2 + \left(\frac{n\pi}{b} \right)^2$$

$m = 1, 2, 3, \dots$ $n = 1, 2, 3, \dots$

When $a \approx b$ $\omega_{1,1} \ll \omega_{1,2}$
(Distinct fundamental mode)

When $a \ll b$ $\omega_{1,1} \approx \omega_{1,2} \approx \omega_{1,3} \approx \dots$
(Many modes are almost degenerate)

Modes have different transverse structure:

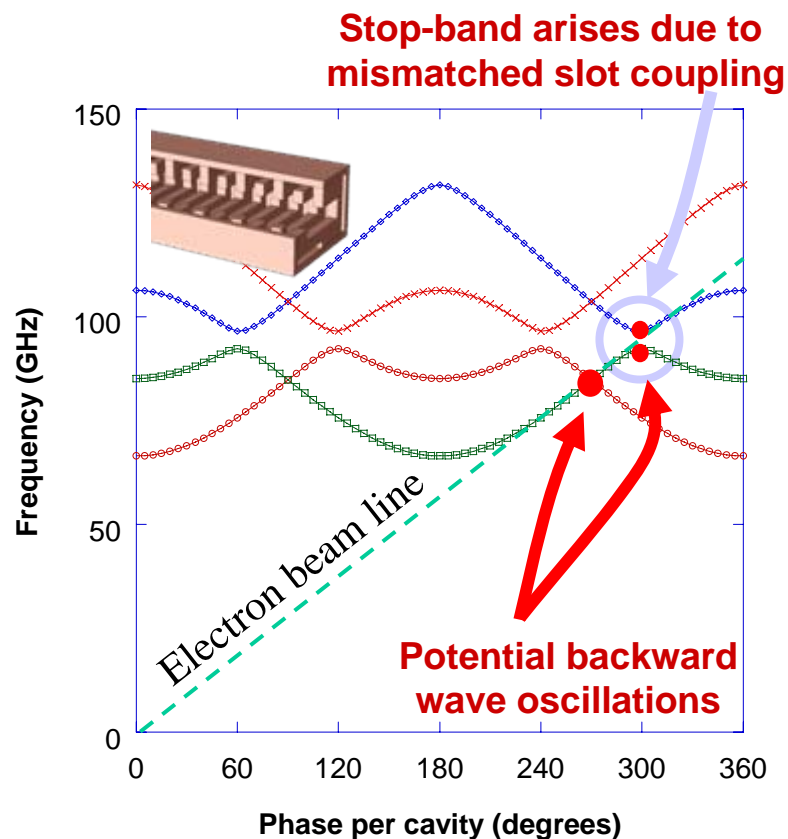
$$E_z \propto \sin \frac{\pi x}{b}, \sin \frac{2\pi x}{b}, \sin \frac{3\pi x}{b}, \dots$$



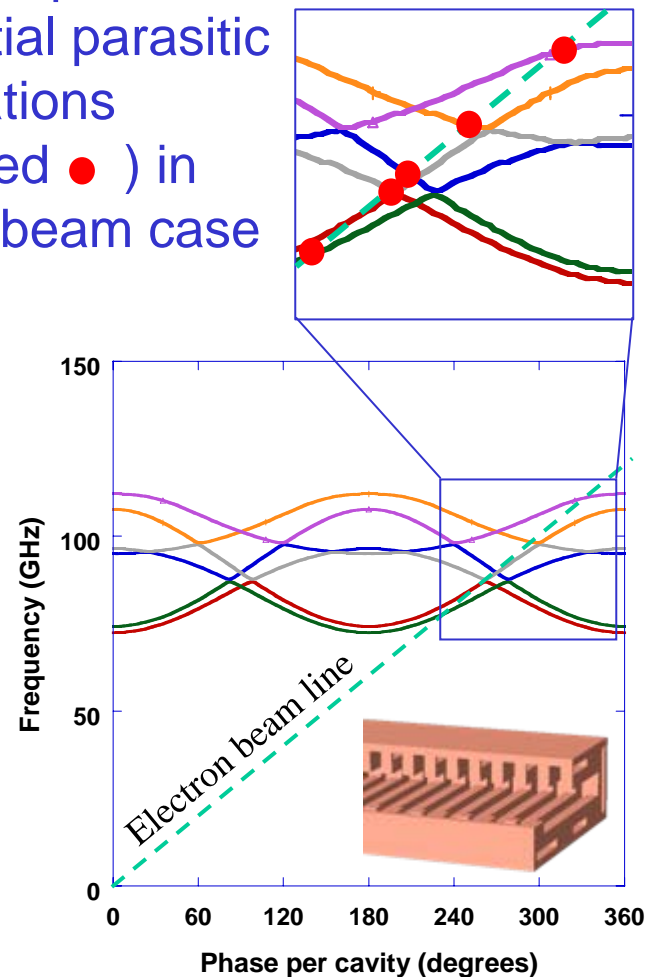
Sources of Mode Competition



- Competing interactions:
 - Backward wave oscillation (BWO)
 - Stop-band oscillations
 - Both TM and TE modes interact



Dense spectrum of potential parasitic oscillations (marked ●) in sheet beam case

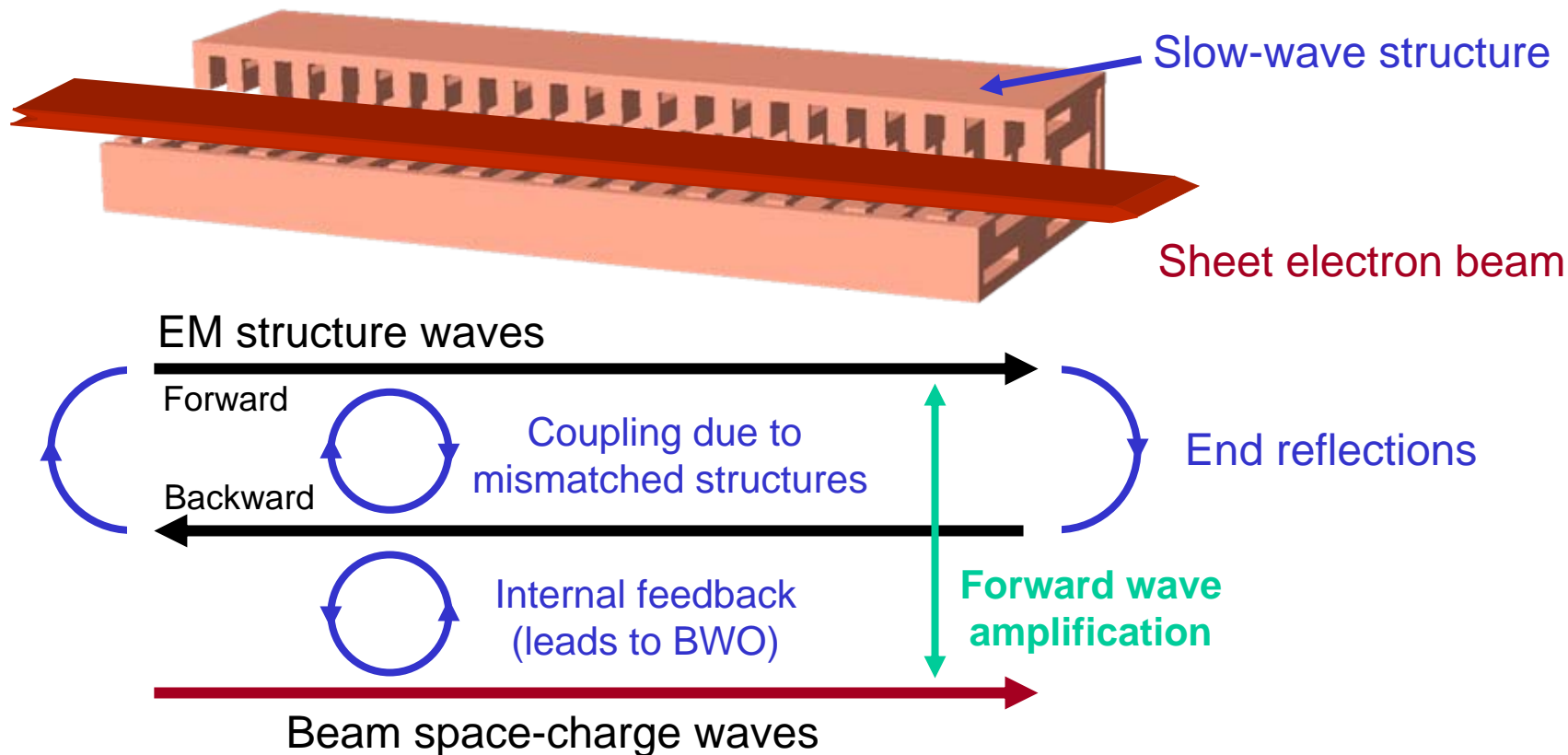




Physical Processes Affecting Stability



Many processes provide feedback, potentially causing oscillation



- BWO oscillation occurs when the gain region exceeds a threshold length
- Stop-bands can significantly decrease this critical length